

面元法电流: $\oint \vec{n} d\vec{l} = \sum I_i$. 平面上: $\vec{i} = \vec{n} \times (\vec{A}^1 - \vec{A}^2)$, $\vec{M} = n I a \vec{S}_a$

$$\vec{i}_0 = \vec{n} \times \frac{1}{\mu_0} (\vec{B}^1 - \vec{B}^2) \quad \vec{i}_0 = \vec{n} \times (\vec{H}^1 - \vec{H}^2)$$

例: $\vec{E} = \vec{B} = ?$ 均匀磁化棒 M 已知

$$\vec{M} = M \hat{z} \quad \vec{S}_a = \pm \hat{z} \quad \text{底面: } \vec{i} = \vec{M} \times \vec{n} = 0 \quad \text{侧: } \vec{i} = \frac{1}{2} \mu_0 \vec{i}' = \frac{1}{2} \mu_0 M \quad (\text{半无限长})$$

已知 \vec{M} , 有: $\vec{M} \rightarrow \vec{i}' \rightarrow \vec{B}$

$$\text{顶面: } \vec{i} = \vec{M} \times \hat{r} = M \hat{\theta} \quad \text{中: } \vec{B} = \mu_0 \vec{i}' = \mu_0 M \quad (\text{无限长})$$

均匀磁化的 $B = \frac{2\mu_0 M}{3}$

$$\oint \vec{H} d\vec{l} = \sum I_0 \quad \vec{H} = \frac{\vec{B}}{\mu_0} - \vec{M}$$

$$\vec{B} = \mu_0 (\vec{H} + \vec{M})$$

磁化率: χ_m , $\vec{M} = \chi_m \vec{H}$

$$\vec{B} = \mu_0 (1 + \chi_m) \vec{H} = \mu_0 \mu_r \vec{H}$$

$\mu_r = 1 + \chi_m$ 相对磁导率

例: $\vec{i} = \vec{M} \rightarrow \vec{B}_1 = 0$

$$\vec{B} = \mu_0 (\vec{H} + \vec{M}) \quad \vec{B} = \mu_0 \mu_r \vec{H}$$

$$\vec{M} = (\mu_r - 1) \vec{H}$$

$$H_1 = \frac{B_1}{\mu_0 \mu_r} = \frac{M}{\mu_r} \quad B_1 = \mu_0 M$$

$$H_2 = \frac{B_2}{\mu_0} = 0 \quad B_2 = 0$$

$$B_2 = \frac{1}{2} \mu_0 M$$

类比: 电动势 \mathcal{E} — 磁动势 $\mathcal{E}_m = NI_0$

电流 I — 磁通量 $\Phi = \iint \vec{B} d\vec{S}$

电阻 R — 磁阻 $R_m = \frac{l}{\mu_0 \mu_r S}$

磁路 $\mathcal{E} = \iint \vec{B} d\vec{S}$
 $\mathcal{E}_m = NI_0$
 $R_m = \frac{l}{\mu_0 \mu_r S}$

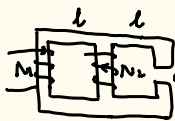


磁通 Φ



$$R_{m1} = \frac{2l-d}{\mu_0 \mu_r S} \quad R_{m2} = \frac{d}{\mu_0 S} \quad (\mu_r = 1)$$

$$\mathcal{E}_m = NI_0 = \frac{\Phi S}{\mu_0} (R_{m1} + R_{m2}) \quad \Phi \text{ 可求}$$



$$R_{m1} = \frac{3l}{\mu_0 \mu_r S} \quad R_{m2} = \frac{l}{\mu_0 \mu_r S} \quad R_{m3} = \frac{2l-d}{\mu_0 \mu_r S} \quad R_{m4} = \frac{d}{\mu_0 S}$$

基本方程: $\begin{cases} \nabla \cdot \vec{B} = 0 \\ \nabla \times \vec{H} = \vec{j} \\ \vec{B} = \mu_0 \mu_r \vec{H} \end{cases}$ 介质的边界条件:

① 各向同性均匀的介质的磁体空间.

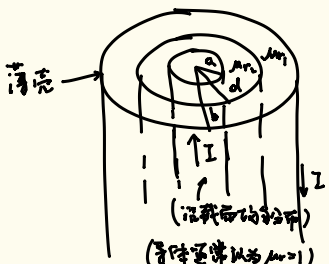
$$\oint \vec{B} d\vec{l} = \begin{cases} \mu_0 I \quad (\text{总电流}) \\ \mu_0 \mu_r \oint \vec{H} d\vec{l} = \mu_0 \mu_r I_0 \quad (\text{传导电流}) \end{cases}$$

$$\Rightarrow B = \mu_r B_0$$

② 不同介质的界面平行于 B

$$\oint \vec{H} d\vec{l} = I_0$$

$$\oint \vec{B} d\vec{l} = \mu_0 I \Rightarrow \text{若 } \vec{i}'_n = 0, \text{ 有 } H = \frac{B}{\mu} \quad (\text{此时对壳})$$



$\vec{B}, \vec{H}, \vec{i}_0, \vec{i}'$?

$$\oint \vec{H} d\vec{l} = I_0$$

$$H_{2\pi r} = \begin{cases} I & r > b \\ 0 & a < r < b \\ \frac{I}{a} & r < a \end{cases} \Rightarrow H = \begin{cases} \frac{I}{2\pi r} & r > b \\ \frac{I}{2\pi a} & a < r < b \\ \frac{I}{2\pi a^2} r & r < a \end{cases}$$

$$M = \mu_r (\mu_r - 1) H = \begin{cases} 0 & r > b \\ (\mu_r - 1) \frac{I}{2\pi r} & a < r < b \\ (\mu_r - 1) \frac{I}{2\pi r} & a < r < b \\ 0 & r < a \end{cases}$$

$$\vec{i}_0 = \hat{n} \times (\vec{H}^1 - \vec{H}^2) \quad (\hat{n} = \hat{r})$$

$$\vec{i}' = \hat{n} \times (\vec{M}^1 - \vec{M}^2)$$

$$\textcircled{1} r = a \quad H^1 = \frac{I}{2\pi a} \hat{\theta}, \quad H^2 = \frac{I}{2\pi a} \hat{\theta} \quad \vec{i}_0 = 0$$

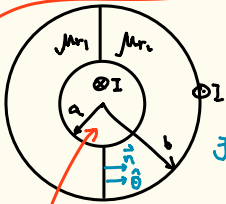
$$M^1 = (\mu_r - 1) \frac{I}{2\pi r} \quad M^2 = 0 \quad \vec{i}' = (\mu_r - 1) \frac{I}{2\pi r} \hat{\theta}$$

② $r = d$ 同轴

$$\textcircled{3} r = b \quad H^1 = 0, \quad H^2 = \frac{I}{2\pi b} \hat{\theta} \quad \vec{i}_0 = -\frac{I}{2\pi b} \hat{\theta} \quad \vec{i}' \text{ 同轴}$$

$$\hat{r} \times \hat{\theta} = \hat{z}$$

② 不同介质, 界面垂直于B, 且通有电流



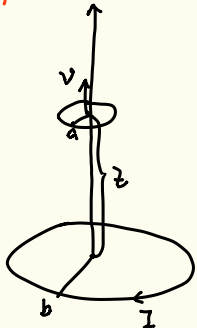
$$\oint \vec{H} \cdot d\vec{l} = \pi r (\mu_1 j_1 + \mu_2 j_2) = I \Rightarrow B = \frac{\mu_1 \mu_2 \mu_0 I}{(\mu_1 + \mu_2) \pi r}$$

于是 $\vec{H} \times \vec{B} = 0$ (界面无电流)

B 连续, 不会突变

一定不是同心!!! B 在界面一定为 0!

电介质情况:



$z \gg b \gg a$

① a 中电势为 ϵ_a ?

② a 处 \vec{E} 与 F_a ?

③ $I_a = \frac{\epsilon_a}{R}, M_a = I_a \cdot \pi a^2$

$F_a = m_a \cdot \frac{\partial B_a}{\partial z}$

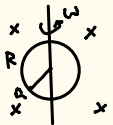
① $\epsilon_a = -\frac{d\phi}{dz}$ (洛伦兹原始公式)

$B_a = \frac{\mu_0 I}{2} \cdot \frac{b^2}{(b^2 + z^2)^{3/2}} \approx \frac{\mu_0 I}{2} \cdot \frac{b^2}{z^3}$

$\epsilon_a = B_a \cdot \pi a^2$

$\Rightarrow \epsilon_a = -\frac{d\phi}{dz} = -\frac{d\phi}{dz} \cdot \frac{dz}{dt} = \frac{3}{2} \frac{\mu_0 I a^2 b^2}{z^4} \cdot v$

对反导体: $\epsilon = \int_a^b \vec{v} \times \vec{B} \cdot d\vec{l}$, $U_{ab} = -\epsilon_{ab}$ 对开路成立



t=0 时与 B 垂直

① ϵ ? I ?

② 保持 ω 旋转时外力矩?

(1) $\epsilon = B \cdot \pi a^2 \cos \omega t \Rightarrow \epsilon = -\frac{d\phi}{dt} = \pi a^2 B \omega \sin \omega t$

$I = \frac{\epsilon}{R} = \frac{\pi a^2 B \omega}{R} \sin \omega t$

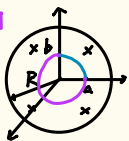
(2) $M = I \cdot \pi a^2 = \frac{\pi^2 a^4 B \omega}{R} \sin^2 \omega t$ (M 与 $\vec{\omega}$ 方向一致)

相同向 $M = \mu B \sin \omega t$

$M_{eff} = M = \frac{\pi^2 a^4 B^2 \omega}{R} \sin^2 \omega t$ 方向相反

或: $P_{\epsilon} = \frac{\epsilon^2}{R} = M_{eff} \cdot \omega$ (外力不做功, 电源功率与外力功率相同)

U 法



$\frac{dB}{dt} = k$

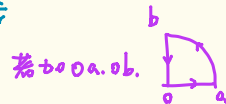
① E 环感?

② 求 U_{ab} ① 导线等一匝线圈 (导线为圆环)

② 无导线

① $\oint \vec{E} \cdot d\vec{l} = -\frac{d\phi}{dt}$ 即 $E \cdot 2\pi r = \begin{cases} r < R: -\frac{d}{dt}(2\pi r^2 B) \\ r > R: -\frac{d}{dt}(2\pi R^2 B) \end{cases}$ 于是 $E = \begin{cases} r < R: -\frac{r}{2} \frac{dB}{dt} = -\frac{k r^2}{2} \\ r > R: -\frac{R^2}{2r} \frac{dB}{dt} = -\frac{k R^2}{2r} \end{cases}$

② $U_{ab} = -\epsilon_{a \rightarrow b}$
 $\epsilon_{a \rightarrow b} = \int_a^b \vec{E} \cdot d\vec{l} = E \cdot 2\pi r_0 = +\frac{1}{2} \pi r_0^2 k$. 则 $U_{ab} = -\epsilon_{a \rightarrow b} = -\frac{1}{2} \pi r_0^2 k$.
 积分方向相反



若加 0a, 0b. 由于 0a, 0b 与 E 垂直, 于是此闭合回路电势差就是 ab 间电势差

正负与方向是重要的

$U_{ab} = -\epsilon_{a \rightarrow b}$
 $\epsilon_{a \rightarrow b} = \int_a^b \vec{E} \cdot d\vec{l} = \frac{1}{2} \cdot 2\pi r_0 \cdot \frac{1}{2} k = \frac{1}{2} \pi r_0^2 k$. 于是有 $U_{ab} = -\epsilon_{a \rightarrow b} = \frac{1}{2} \pi r_0^2 k$

③ $U_{ab} = 0$ (两空同无电势差! 电势差一定沿路径)

如下题:



$\frac{dB}{dt} = k, U_{AB} = ?$

环绕方向!!

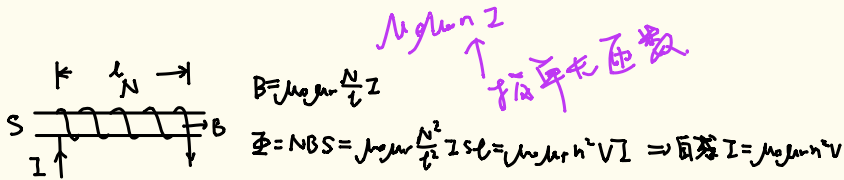
$\epsilon_{A \rightarrow B} = \epsilon_{ABOA} = -\frac{d\phi}{dt} = -\frac{d}{dt}(\frac{1}{2} \pi R^2 B) = \frac{1}{2} \pi k R^2$

$U_{AB} = -\epsilon_{A \rightarrow B} = -\frac{1}{2} \pi k R^2$

$U_{AB} = -\epsilon_{A \rightarrow B} = \epsilon_{B \rightarrow A}$

$\epsilon_{B \rightarrow A} = \epsilon_{B \rightarrow O \rightarrow A} = -\frac{d\phi}{dt} = \frac{1}{2} \pi R^2 k$ 或: $\epsilon_{O \rightarrow B \rightarrow A} = \frac{1}{2} \pi k R^2$

$\epsilon_{O \rightarrow B \rightarrow A} = \frac{1}{2} \pi k R^2 = \epsilon_{B \rightarrow A} = -\epsilon_{A \rightarrow B} = U_{AB}$



$\mu_0 \mu_r n^2$
 ↑
 电感系数



单位长度自感 L?
 ① 高频近似 (壳面近似!)
 ② 轴对称电路

$\omega_n = \frac{1}{2} \nabla \cdot \vec{H}$
 $H = \begin{cases} \frac{r}{2a} I & r < R_1 \\ \frac{I}{r} & R_1 < r < R_2 \\ \frac{I}{r} & r > R_2 \end{cases} \Rightarrow \omega = \begin{cases} \frac{\mu_0 I^2}{8\pi^2 R_1^2} & r < R_1 \\ \frac{\mu_0 I^2}{8\pi^2 r^2} & R_1 < r < R_2 \\ \frac{\mu_0 I^2}{8\pi^2 r^2} & r > R_2 \end{cases}$

$\int_{R_1}^{R_2} \omega dr$ 为 $\omega_m \cdot 2\pi r dr$
 $\omega_m = \int_0^{R_1} \omega dr + \int_{R_1}^{R_2} \omega dr$
 $= \frac{\mu_0 I^2}{16\pi} + \frac{\mu_0 \mu_r I^2}{4\pi} \ln \frac{R_2}{R_1}$

$\frac{\mu_0 \mu_r}{16\pi R_1^2} I^2 \quad L = \frac{2\omega}{I^2}$

$\oint \vec{H} \cdot d\vec{l} = I_{in} = \begin{cases} 0 & r < R_1 \\ I & R_1 < r < R_2 \\ 0 & r > R_2 \end{cases}$ 于是 $H = \begin{cases} \frac{I}{2\pi r} & R_1 < r < R_2 \\ 0 & \text{others} \end{cases}$
 $B = \begin{cases} \frac{\mu_0 \mu_r I}{2\pi r} & R_1 < r < R_2 \\ 0 & \text{others} \end{cases}$
 $\Rightarrow d\omega = \omega dr = \omega dr$ 于是 $\omega = \int_{R_1}^{R_2} \frac{\mu_0 \mu_r I^2}{2\pi r} dr = \frac{\mu_0 \mu_r I^2}{2\pi} \ln \frac{R_2}{R_1} \Rightarrow \text{自感 } L = \frac{\omega}{I^2} = \frac{\mu_0 \mu_r}{2\pi} \ln \frac{R_2}{R_1}$

不是 $2\pi dr$! 看 \vec{H} 方向!

$\oint \vec{H} \cdot d\vec{l} = I_{in} = \begin{cases} I \frac{r}{R_1} & r < R_1 \\ I & R_1 < r < R_2 \\ 0 & r > R_2 \end{cases}$
 $B = \begin{cases} \frac{\mu_0 \mu_r r}{2\pi R_1^2} I & r < R_1 \\ \frac{\mu_0 \mu_r I}{2\pi r} & R_1 < r < R_2 \\ 0 & r > R_2 \end{cases}$
 $\Rightarrow \omega = \int_0^{R_1} B \omega r + \int_{R_1}^{R_2} B \omega r = \frac{\mu_0 \mu_r I^2}{4\pi} + \frac{\mu_0 \mu_r I^2}{2\pi} \ln \frac{R_2}{R_1} \Rightarrow L = \frac{\omega}{I^2} = \frac{\mu_0 \mu_r}{2\pi} \ln \frac{R_2}{R_1} + \frac{\mu_0 \mu_r}{4\pi}$

$\omega = \int_0^{R_1} B \frac{I}{2\pi} dr \quad I = \frac{I}{R_1} r$

此部分不能直接除 I, 是错误!

理想磁芯时自感 $L = \sqrt{L_1 L_2}$

串联顺接: $L = L_1 + L_2$ 同向: $L = L_1 + L_2 - M = (\sqrt{L_1} - \sqrt{L_2})^2$

并联异向: $L = \frac{L_1 L_2 - M^2}{L_1 + L_2 - 2M}$ 同向: $L = \frac{L_1 L_2 - M^2}{L_1 + L_2 + 2M}$

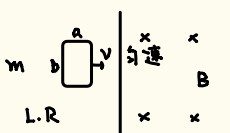
特例: $L_1 = L_2 = L_0$ 时理想情况下仍为 L_0 .

R-C 电路: $i = I_0 (1 - e^{-\frac{t}{\tau}})$, $\tau = RC$

故: $i = I_0 e^{-\frac{t}{\tau}}$

R-L 电路: $i = I_0 (1 - e^{-\frac{t}{\tau}})$, $\tau = \frac{L}{R}$

故: $i = I_0 e^{-\frac{t}{\tau}}$



水平无磁感, $t=0$ 时开进

$i(t) = ?$

$\tau_L = \frac{L}{R}$

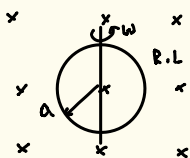
$i = i_0 \cdot e^{-\frac{t}{\tau}}$

不要弄成 t_0

$\text{电压: } \varepsilon = B l v$
 $i = i_0 (1 - e^{-\frac{t}{\tau}})$
 $i_0 = \frac{\varepsilon}{R}$
 $t_0 = \frac{L}{R}$
 $(0 < t < t_0)$

$\text{磁通: } \Phi = \Phi_0 (1 - e^{-\frac{t}{\tau}})$
 $\dot{\Phi} = \dot{\Phi}_0 (1 - e^{-\frac{t}{\tau}})$

$\varepsilon = RI + L \frac{dI}{dt} + \frac{1}{C} \int I dt$



$i(t) = ?$

$\begin{cases} \varepsilon = iR + L \frac{di}{dt} \\ \varepsilon = -\frac{d\Phi}{dt} = B 2a^2 \omega \sin \omega t \end{cases}$

电流取 Re, 阻抗取 |·|

$\tilde{\varepsilon} = \varepsilon_0 e^{j(\omega t + \varphi_0)}$ $\varphi_0 = -\frac{\pi}{2}$, $\varepsilon_0 = B 2a^2 \omega$

$\tilde{\varepsilon} = R + j\omega L = \sqrt{R^2 + \omega^2 L^2} e^{j\varphi}$, $\tan \varphi = \frac{\omega L}{R}$

$\tilde{I} = \frac{\tilde{\varepsilon}}{Z} = \frac{\varepsilon_0}{\sqrt{R^2 + \omega^2 L^2}} e^{j(\omega t + \varphi_0 - \varphi)}$

$\Rightarrow i(t) = \text{Re}(\tilde{I}) = \frac{\varepsilon_0}{\sqrt{R^2 + \omega^2 L^2}} \cos(\omega t - \frac{\pi}{2} - \varphi)$

有效值: $\frac{1}{\sqrt{2}} \tilde{I}$

交流电路: $\tilde{U} = U_m e^{j(\omega t + \varphi_0)}$ \tilde{I} , $\tilde{\varepsilon}$ 同理. $\tilde{U} = \tilde{I} Z$ Z 为阻抗, 其中 $\tilde{Z}_R = R$, $\tilde{Z}_C = \frac{1}{j\omega C}$, $\tilde{Z}_L = j\omega L$. 阻抗取模即为阻抗

L 电路: $L \frac{dI}{dt}$



$t=0$ 时磁通 $\Phi = \frac{Q}{\epsilon} + L \frac{dQ}{dt} \quad (\Phi = \int I dt + L \frac{dQ}{dt})$

- (1) L中磁通量 = C中电荷量 Q_1 (等-1k)
- (2) L中磁通量 = 电容电压 U_2

\hookrightarrow (1) $\frac{Q}{\epsilon} + L \frac{dQ}{dt} = 0 \quad i = \frac{dQ}{dt}$

$Q = Q_0 \cos(\omega t + \phi) \quad \omega_0 = \frac{1}{\sqrt{LC}}$

$t=0$ 时 $Q=Q_0, \quad i=0 \Rightarrow Q = Q_0 \cos \omega t$

$i = -\omega Q_0 \sin \omega t$

(2) $\omega_0 t = \frac{\pi}{2} \Rightarrow t = \frac{\pi}{2\omega_0}$

$W_e = \frac{1}{2} \frac{Q^2}{\epsilon} = \frac{1}{2} \frac{Q_0^2}{\epsilon} \cos^2 \omega t$

$W_m = \frac{1}{2} L i^2 = \frac{1}{2} \frac{L Q_0^2}{\epsilon^2} \sin^2 \omega t$

$\Rightarrow t_1 = \frac{\pi}{2\omega_0}$

两线圈磁通: $\Phi_{12} = \frac{1}{2} L_1 I_1^2 + \frac{1}{2} L_2 I_2^2 + M I_1 I_2$

$W_m = \frac{1}{2} \frac{L_1}{\mu_0} I_1^2$; Φ_1 为第1个线圈的磁通

互感磁通 $W = \Phi_{12} I_2 = M I_1 I_2$ 只考虑互感磁通(线圈磁通和不在 I_1, I_2 不在)

对均匀外场 $W = \int \vec{J} \cdot \vec{a} = \vec{m} \cdot \vec{b}$

$\oint \vec{H} \cdot d\vec{l} = I_{in}$

$\Rightarrow \int 2\pi r H = \begin{cases} \frac{r^2}{a^2} I & r < a \\ I & a < r < b \\ I - \frac{r^2 - b^2}{c^2 - b^2} I = \frac{c^2 - r^2}{c^2 - b^2} I & b < r < c \\ 0 & r > c \end{cases}$ $\text{Mag } H = \begin{cases} \frac{r}{2a^2} I & r < a \\ \frac{I}{2\pi r} & a < r < b \\ \frac{c^2 - r^2}{c^2 - b^2} \cdot \frac{I}{2\pi r} & b < r < c \end{cases}$

① $r < a$

$W_{m1} = \frac{1}{2} \int \vec{B} \cdot \vec{H} = \frac{1}{2} \int \mu_0 H^2 = \frac{\mu_0 I^2}{8\pi^2 a^2} \int_0^a 2\pi r \cdot W_{m1} dr = \frac{\mu_0 I^2}{16a}$

② $a < r < b$

$W_{m2} = \frac{1}{2} \int \vec{B} \cdot \vec{H} = \frac{1}{2} \int \mu_0 H^2 = \frac{\mu_0 I^2}{8\pi^2 r^2} \int_a^b 2\pi r \cdot W_{m2} dr = \frac{\mu_0 I^2}{4\pi} \ln \frac{b}{a}$

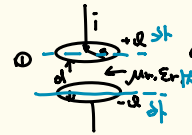
③ $b < r < c$

同理得 $W_{m3} = \frac{\mu_0 I^2}{4\pi(c^2 - b^2)} (c^2 \ln \frac{c}{b} - c^2 + c^2 \frac{b^2}{c^2} + \frac{1}{2}(c^2 - b^2))$

$W = W_{m1} + W_{m2} + W_{m3} = \dots = \frac{1}{2} L I^2$ 则 $L = \frac{2W}{I^2}$ (结果不用算)

直接: 磁场的做功与磁场的增加! 电池做功等于磁场的增加

- 计算位移电流密度 $\frac{\partial D}{\partial t}$
- ① $E_{ext} \leftarrow \oint \vec{D} \cdot d\vec{s} = q_0$
 - ② $E_{ind} \leftarrow \oint \vec{E} \cdot d\vec{l} = -\frac{d\Phi}{dt}$
 - ③ $\vec{j} = \sigma \vec{E}$



1° 求位移电流密度 \vec{j}_D
2° 求磁通 B 分布 (忽略 B 对 E 的影响)

1° 内 $r < a, \quad D = \sigma_0 = \frac{Q}{2a^2} = \frac{Q_0}{2a^2} \cos \omega t, \quad \vec{j}_D = \frac{\partial D}{\partial t} = -\omega \frac{Q_0}{2a^2} \sin \omega t$
其他 $\vec{j}_D = 0$

2° 内 $\oint \vec{H} \cdot d\vec{l} = \int (\vec{j} + \frac{\partial D}{\partial t}) \cdot d\vec{s}$ (此题恒有 $\vec{j} = 0$)

$\Rightarrow \int 2\pi r H = \begin{cases} -\omega \frac{Q_0}{2a^2} \sin \omega t \cdot \pi r^2 & r < a \\ -\omega \frac{Q_0}{2a^2} \sin \omega t \cdot \pi a^2 & r > a \end{cases} \Rightarrow H = \begin{cases} -\frac{\omega Q_0}{2a^2} \sin \omega t & r < a \\ -\frac{\omega Q_0}{2a^2} \sin \omega t & r > a \end{cases} \quad B = \mu_0 H = \begin{cases} \mu_0 H & r < a \\ \mu_0 H & r > a \end{cases}$

外 $\oint \vec{H} \cdot d\vec{l} = i = \frac{dQ}{dt}$

$\int 2\pi r H = -\omega Q_0 \sin \omega t \quad H$ 可求 B 分布

③ $B = B_0 \cos(\omega t)$ (忽略 E 对 B 的影响) (或相线论 \vec{j}_D) $I = I_0 \cos \omega t$
求位移电流密度 \vec{j}_D $B = \mu_0 j_D r$

正方向为 $t=0$ 或 $t=0$ 时 \vec{r} 与 B 同方向

$\oint \vec{E} \cdot d\vec{l} = E \cdot 2\pi r = \begin{cases} \pi r^2 \frac{dD}{dt} & r < a \\ \pi a^2 \frac{dD}{dt} & r > a \end{cases} = \begin{cases} \frac{1}{2} B_0 \omega \sin \omega t & r < a \\ \frac{a^2}{2r} B_0 \omega \sin \omega t & r > a \end{cases} \quad D = \epsilon_0 E \Rightarrow \vec{j}_D = \frac{\partial D}{\partial t} = \begin{cases} \epsilon_0 \frac{1}{2} B_0 \omega^2 \cos \omega t & r < a \\ \epsilon_0 \frac{a^2}{2r} B_0 \omega^2 \cos \omega t & r > a \end{cases}$

④ $I_0 = I_0 \cos \omega t$

$\vec{j}_D = \sigma \vec{E}$
 $\vec{j}_D = \frac{I_0}{\pi} = \frac{I_0}{\pi} \cos \omega t \Rightarrow E = \frac{1}{\sigma} \vec{j}_D = \frac{I_0}{\sigma} \cos \omega t$ 则 $D = \epsilon_0 \epsilon_r E = \frac{\epsilon_0 \epsilon_r}{\sigma} \cdot \frac{I_0}{\pi} \cos \omega t$

$\vec{j}_D = \frac{\partial D}{\partial t} = -\omega \frac{\epsilon_0 \epsilon_r}{\sigma} \cdot \frac{I_0}{\pi} \sin \omega t$

$i_D = \vec{j}_D \cdot \vec{s} = -\frac{I_0 \epsilon_0 \epsilon_r \omega}{\sigma} \sin \omega t$



$t=0$ 时电荷分布为 Q_0

t 时电荷为 $Q_0 e^{-\frac{r}{\lambda} t}$

① $\oint \vec{D} \cdot d\vec{s} = q_0 \Rightarrow Ds = q_0$

$\oint \vec{J}_d \cdot d\vec{s} = -\frac{dq_0}{dt} \Rightarrow \vec{J}_d = -\frac{dq_0}{dt} \cdot \frac{1}{s} \hat{s}$
 $\vec{J}_d = \sigma \vec{E} \Rightarrow \sigma E = \epsilon_0 \epsilon_r \frac{dE}{dt}$

同理 $E = E_0 e^{-\frac{r}{\lambda} t}$

$E_0 = \frac{Q_0}{4\pi \epsilon_0 \epsilon_r r^2} \Rightarrow \vec{J}_d = \sigma E = \frac{\sigma Q_0}{4\pi \epsilon_0 \epsilon_r r^2} e^{-\frac{r}{\lambda} t}$

② $\vec{J}_0 = \frac{\partial \vec{D}}{\partial t} = \epsilon_0 \epsilon_r \frac{\partial \vec{E}}{\partial t} = -\frac{Q_0 \sigma}{4\pi \epsilon_0 \epsilon_r r^2} e^{-\frac{r}{\lambda} t} \hat{s}$

于是 \vec{J}_t (总电流) $= \vec{J}_0 + \vec{J}_d = 0$ 无净电流, $B=0$.

Maxwell 方程组:
$$\begin{cases} \oint \vec{D} \cdot d\vec{s} = \Sigma q \\ \oint \vec{E} \cdot d\vec{l} = -\int \frac{\partial \vec{B}}{\partial t} \cdot d\vec{s} \\ \oint \vec{B} \cdot d\vec{s} = 0 \\ \oint \vec{H} \cdot d\vec{l} = \Sigma I + \int \frac{\partial \vec{D}}{\partial t} \cdot d\vec{s} \end{cases} \quad \begin{cases} \nabla \cdot \vec{D} = \rho \\ \nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \\ \nabla \cdot \vec{B} = 0 \\ \nabla \times \vec{H} = \vec{j} + \frac{\partial \vec{D}}{\partial t} \end{cases}$$

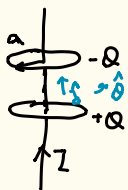
能量流密度 $\vec{S} = \vec{E} \times \vec{H}$



电压 $U = \frac{Q}{C}$ 求 $\vec{S} = ?$

有 $E \propto \frac{1}{r}$, 且 $E = \frac{Q}{C} \Rightarrow E = \int_a^b \frac{Q}{4\pi r^2} dr = C \left(\frac{1}{a} - \frac{1}{b} \right) \Rightarrow C = \frac{E}{\frac{1}{a} - \frac{1}{b}}$ 于是 $E = \frac{U}{\frac{1}{a} - \frac{1}{b}}$
 $\oint \vec{H} \cdot d\vec{l} = I \Rightarrow \vec{H} = \frac{I}{2\pi r} \hat{\phi} \Rightarrow \vec{S} = \vec{E} \times \vec{H} = \frac{E}{\frac{1}{a} - \frac{1}{b}} \cdot \frac{I}{2\pi r} (\hat{r} \times \hat{\phi}) = \frac{EI}{2\pi r^2 \left(\frac{1}{a} - \frac{1}{b} \right)} \hat{z}$

$\int \vec{S} \cdot d\vec{A} = \int_a^b \frac{EI}{2\pi r^2 \left(\frac{1}{a} - \frac{1}{b} \right)} \cdot 2\pi r dr = EI = I^2 R$ (与 [1] 中结论一致)

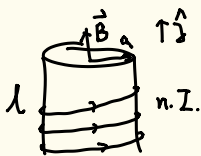


$Q = I t, E = \frac{Q}{\epsilon_0 s} = \frac{I t}{\epsilon_0 2\pi a t} \hat{z}$

$D = \epsilon_0 E \Rightarrow \vec{J}_d = \frac{\partial D}{\partial t} = \frac{I}{2\pi a} \hat{z}$ 于是 $r=a$ 时 $I = 2\pi a J_d$

$\oint \vec{H} \cdot d\vec{l} = (I_0 + I_d)_{enc} = \begin{cases} r < a & J_d \cdot 2\pi r^2 = \frac{I}{2\pi a} \cdot 2\pi r^2 \\ r > a & J_d \cdot 2\pi a^2 = I \end{cases} \quad \vec{H} = \begin{cases} \frac{I}{2\pi a} r \hat{\phi} & r < a \\ \frac{I}{2\pi r} \hat{\phi} & r > a \end{cases}$

于是 $\vec{S} = \vec{E} \times \vec{H} = \begin{cases} \frac{I^2 r}{2\pi \epsilon_0 a^2} \hat{r} & (r < a) \\ 0 & (E=0) \end{cases}$ (与 [1] 中结论一致)



$I = k t \quad \vec{S} ?$

$\vec{B} = \mu_0 n I \hat{z} = k \mu_0 n t \hat{z}$

$\oint \vec{E} \cdot d\vec{l} = -\frac{dB}{dt} \Rightarrow 2\pi r \vec{E} = -\frac{dk}{dt} \pi r^2 \Rightarrow \vec{E} = -\frac{r}{2} k \mu_0 n \hat{\phi}$

$\vec{S} = \vec{E} \times \vec{H} = -\frac{r}{2} k \mu_0 n^2 t \hat{r}$

$\oint \vec{S} \cdot d\vec{A} = -\frac{a}{2} k \mu_0 n^2 t \cdot 2\pi a l$ (与 [1] 中结论一致)

$W_m = \frac{B^2}{2\mu_0} \cdot 2\pi a l \Rightarrow \frac{dW_m}{dt} = \frac{dB}{dt} \cdot \frac{dW_m}{dB}$ (与 [1] 中结论一致)

能量流密度 $\vec{g} = \vec{D} \times \vec{B}$